Antennas and Free Space Propagation

WIRELESS SHORT COURSE

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Learning Objectives

- ❑Mathematically describe an EM wave:
  - Direction of motion, wavenumber, frequency, polarization, …
- ❑Identify radio spectrum and power levels used in common commercial wireless products
- ❑Perform basic power calculations in dB scale
- ❑Perform basic mathematical operations in polar coordinates
  - Conversions to cartesian coordinates, rotations, integrals, averages, …
- ❑Use tools from MATLAB to compute and plot key antenna parameters
  - Directivity, gain, efficiency, …
- ❑Compute received power in an angular region using the radiation density and intensity.
- ❑Compute the free-space path loss using Friis Law
- ❑Derive Friis Law

Outline

- ❑ Basics of Electromagnetic Waves
- ❑ Power and Bandwidth of Signals
- ❑ Basics of Antennas
- ❑ Free Space Propagation

Electric and Magnetic Forces

- Two closely related forces:
  - Electric: Forces between charged particles
  - Magnetic: Forces between moving charged particles
- Forces operate at a distance:
  - Enables communication.
  - … and many other phenomena in the universe
- Represented by a vector field
  - Force strength has a direction and magnitude
  - Changes with position $\boldsymbol{r} = (x, y, z)$ and time t
  - E-field: $\boldsymbol{E}(\boldsymbol{r}, t)$ in N/C (force / unit charge)
  - B-field: $\boldsymbol{B}(\boldsymbol{r}, t)$ in N/(Am) (force / unit charge / velocity)

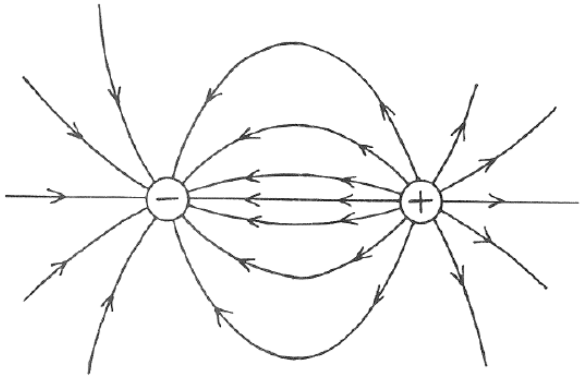
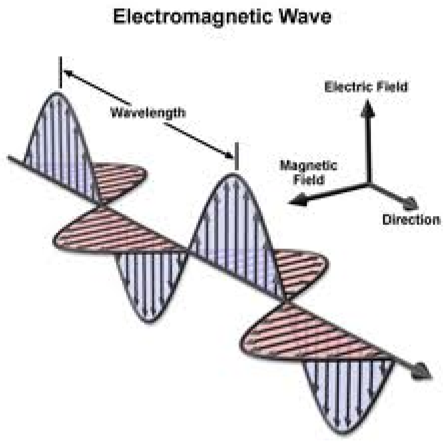


FIGURE 2.4 Electric field lines begin and end on charges.

Plane Waves

- EM field governed by Maxwell's equations
- All solutions can be decomposed into plane waves



- EM plane wave at position $\boldsymbol{r} = (x, y, z)$
  - $\boldsymbol{E}(\boldsymbol{r}, t) = E_0 \boldsymbol{e}_x \cos(2\pi(ft + \lambda^{-1}z) + \phi)$
  - $\boldsymbol{B}(\boldsymbol{r}, t) = B_0 \boldsymbol{e}_y \cos(2\pi(ft + \lambda^{-1}z) + \phi)$
  - $B_0 = (1/c)E_0, \quad c = \lambda f =$ speed of light
- Sometimes write:
  - $\boldsymbol{E}(\boldsymbol{r}, t) = E_0 \boldsymbol{e}_x \cos(\omega t + kz + \phi)$
  - $\boldsymbol{B}(\boldsymbol{r}, t) = B_0 \boldsymbol{e}_y \cos(\omega t + kz + \phi)$
  - $k = \frac{2\pi}{\lambda} =$ wave number

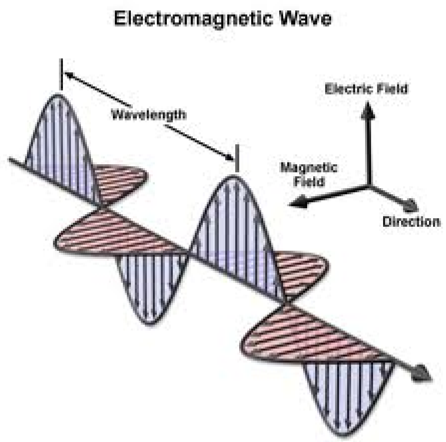
Plane Waves Illustrated

- EM plane wave
  - $\boldsymbol{E}(\boldsymbol{r},t) = E_0\boldsymbol{e}_x \cos(2\pi(ft - \lambda^{-1}z) + \phi)$
  - $\boldsymbol{B}(\boldsymbol{r},t) = B_0\boldsymbol{e}_y \cos(2\pi(ft - \lambda^{-1}z) + \phi)\,, B_0 = c^{-1}E_0$
- Five key parameters:
  - Amplitude, frequency, direction of motion, phase
  - Polarization (see below)
- Diagrams on board
  - Fixed position, variation in time
  - Fixed time, variation in position.
- Phasor notation: $\boldsymbol{E}(\boldsymbol{r},t) = Real\left[\boldsymbol{E}(\boldsymbol{r})e^{i\omega t}\right]$

Plane Wave Direction of Motion

- ❑EM field constant in $x - y$ plane
- ❑Moves along z-direction:

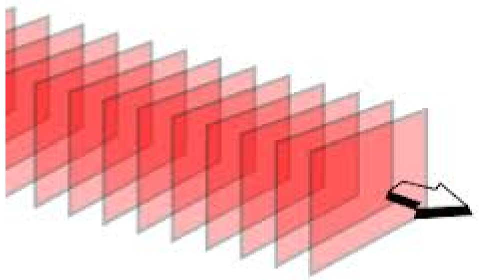
$$E(x, y, z, t + \delta t) = E(x, y, z - c\delta t, t)$$



- ❑"Poynting" vector:

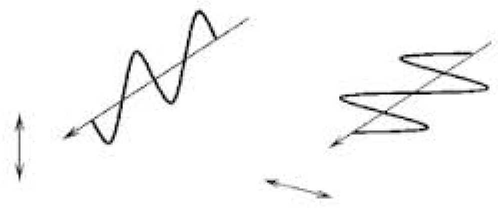
$$\boldsymbol{S} = \frac{1}{\mu}\boldsymbol{E} \times \boldsymbol{B} = \frac{|E_0|^2}{c\mu}\cos^2(2\pi(ft - \lambda^{-1}z))\boldsymbol{e}_z$$

  - Represents "energy flux"
  - Energy consumed = $\nabla \cdot \boldsymbol{S}$
  - Units = W/m^2
  - $\eta = c\mu =$characteristic impedance
  - Vacuum: $\eta = \eta_0 \approx 377\Omega$

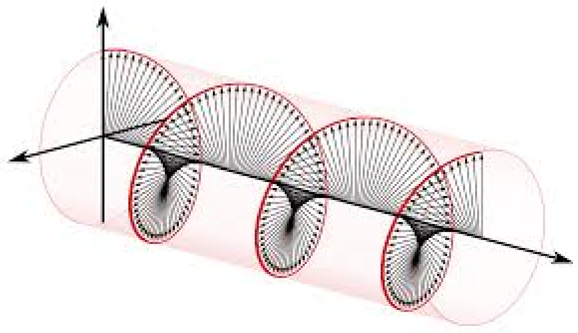


Polarization

- **Polarization**: Orientation of E-field relative to direction of motion
- **Linearly polarized**: Constant orientation
  - Vertical: $\boldsymbol{E}(\boldsymbol{r},t) = E_0\boldsymbol{e}_x\cos(\omega t + kz)$
  - Horizontal: $\boldsymbol{E}(\boldsymbol{r},t) = E_0\boldsymbol{e}_y\cos(\omega t + kz)$



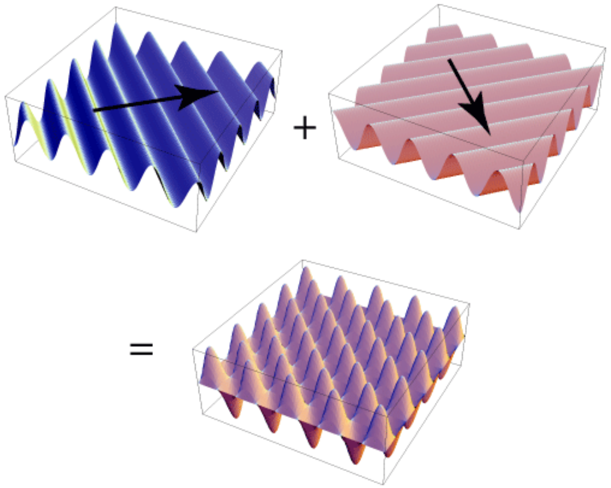
- Also, **circularly polarized**
  - Sum of V and H that are out of phase
  - $E_0[\boldsymbol{e}_x\cos(\omega t + kz) \pm \boldsymbol{e}_y\sin(\omega t + kz)]$
  - Called left hand and right hand



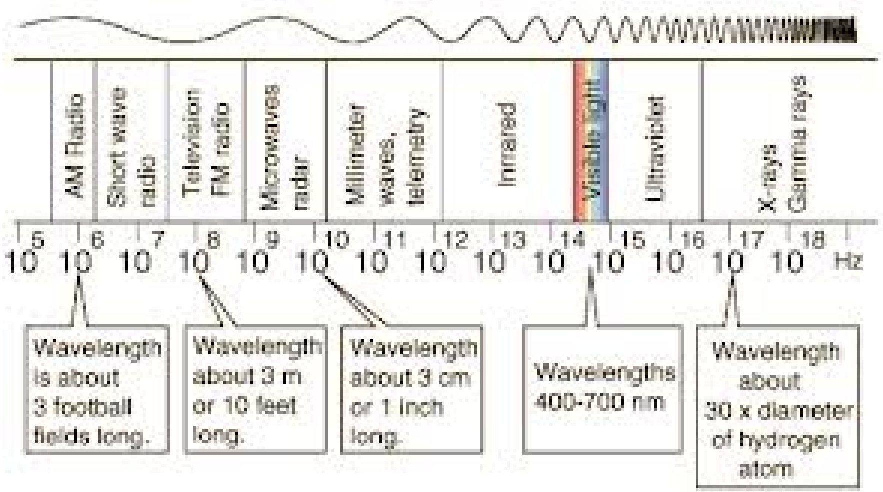
- **Two degrees of freedom**:
  - Consider any plane wave in some direction
  - Can be decomposed as V + H or LH + RH

Plane Wave Decomposition

- Every electric field is a linear combination of plane waves
- Each plane wave in the decomposition has:
  - Frequency
  - Direction of motion
  - Gain, Phase
  - One of two polarization
- Decomposition can be found from a 4D Fourier transform
  - $\boldsymbol{E}(x,y,z,t) \Rightarrow \hat{E}_V(k_x,k_y,k_z,f)$ and $\hat{E}_H(k_x,k_y,k_z,f)$
  - Converts time + space ⇒ wavenumber and frequency
  - Note that there are two polarization components
- This decomposition is used in many EM solvers
  - And your EM class if you take it



EM Spectrum



- ❑Frequency of EM radiation has wide range
- ❑Encompasses many forms of radiation
- ❑Radio waves are uniquely valuable since they can propagate far

Outline

- Basics of Electromagnetic Waves
- Power and Bandwidth of Signals
- Basics of Antennas
- Free Space Propagation

Signals for Communication

Antenna

TX signal
Carries information

EM radiation

- Signal: Any quantity that varies in time
  - Continuous, discrete, complex, real, …
- Signals for wireless communications:
  - Modulate an information bearing signal to a signal in the EM radiation
- Three key characteristics of the signal: power, bandwidth, center frequency

Energy and Power of Signals

- Consider a scalar-valued, continuous-time signal $x(t)$
- Define instantaneous power: $|x(t)|^2$
- Typically $|x(t)|^2$ this is proportional to the actual power
  - Ex 1: For a voltage, power $= \frac{|V(t)|^2}{R}$
  - Ex 2: For an EM plane wave , power flux $= \frac{|E(t)|^2}{\eta}$
- Energy:
  - $E_x = \int_{-\infty}^{\infty} |x(t)|^2 dt$
  - Signal is called an “energy signal” if $E_x < \infty$
- Power:
  - $P_x = \lim_{T \to \infty} \frac{1}{T} \int_{-T/2}^{T/2} |x(t)|^2 dt$
  - Energy per unit time
  - Signal is called a “power signal” if limit P_x exists and is finite

Power: Linear and decibel scale

- Receive or transmit energy per unit time
  - Measured in Watts (W) or mW
  - Power values in W or mW called *linear scale*
  - Use notation $P_{|W}$ or $P_{|mW}$ when units need to be specified

- Power often measured in dB scale:
  - $P_{|dBW} = 10\log_{10}(P_{|W} / 1W)$
  - $P_{|dBm} = 10\log_{10}(P_{|mW} / 1mW)$

- Example: P = 250 mW (typical max mobile transmit power)
  - $P_{|dBW} = 10\log_{10}(0.25W / 1W)$
  - $P_{|dBm} = 10\log_{10}(250mW / 1mW)$

Some important dB values

- Some conversions don't need a calculator:
  - 10log10(2) = 3 [Most important: Doubling power = 3dB]
  - 10log10(3) =4.7 ~5
  - 10log10(10) = 10

- You can cascade these.
- Ex: What is 50 mW in dBm?
- Ans:

$$\begin{aligned} 10\log_{10}(50) &= 10\log_{10}(10^2/2) \\ &= (2)10\log_{10}(10) - 10\log_{10}(2) = 2(10) - 3 = 17 \text{ dBm} \end{aligned}$$

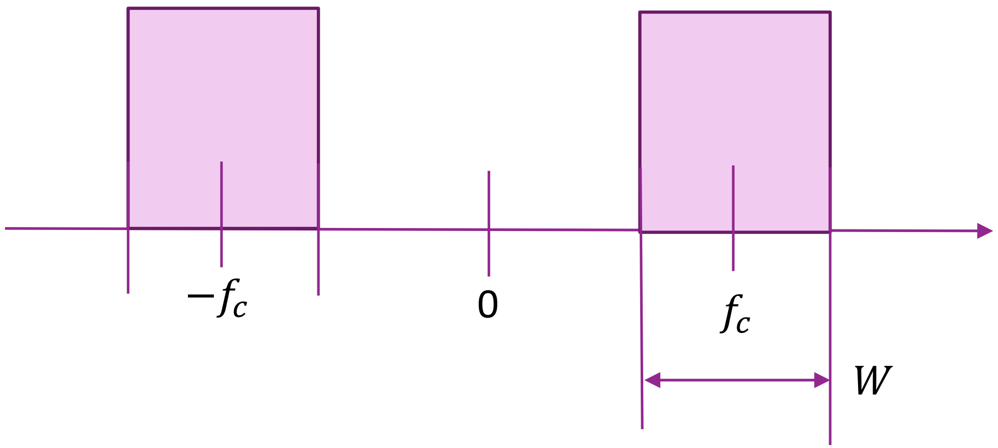
Typical Wireless Power Transmit Levels

- 100 kW = 80 dBm: Typical FM radio transmission with 50 km radius
- 1 kW = 60 dBm: Microwave oven element (most of this doesn't escape)
- ~300 W = 55 dBm: Geostationary satellite
- 250 mW = 24 dBm: Cellular phone maximum power (class 2)
- 200 mW = 23 dBm: WiFi access point
- 32 mW = 15 dBm: WiFi transmitter in a laptop
- 4 mW = 6 dBm: Bluetooth 10 m range
- 1 mW = 0 dBm: Bluetooth, 1 m range

Bandwidth and Carrier Frequency

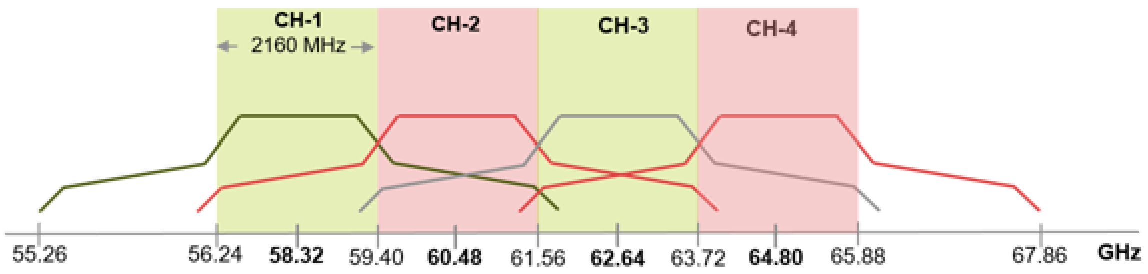
- Power density spectrum:
  - All signals can be decomposed into frequency components (Use Fourier transform)
  - PSD measures the power per unit frequency
  - Indicates range of frequencies of the corresponding EM wave
  - Measured by a spectrum analyzer

- Two key parameters for RF signals:
  - Carrier or center frequency, $f = f_c$
  - Bandwidth W
- Note for a real-valued signal:
  Always two images

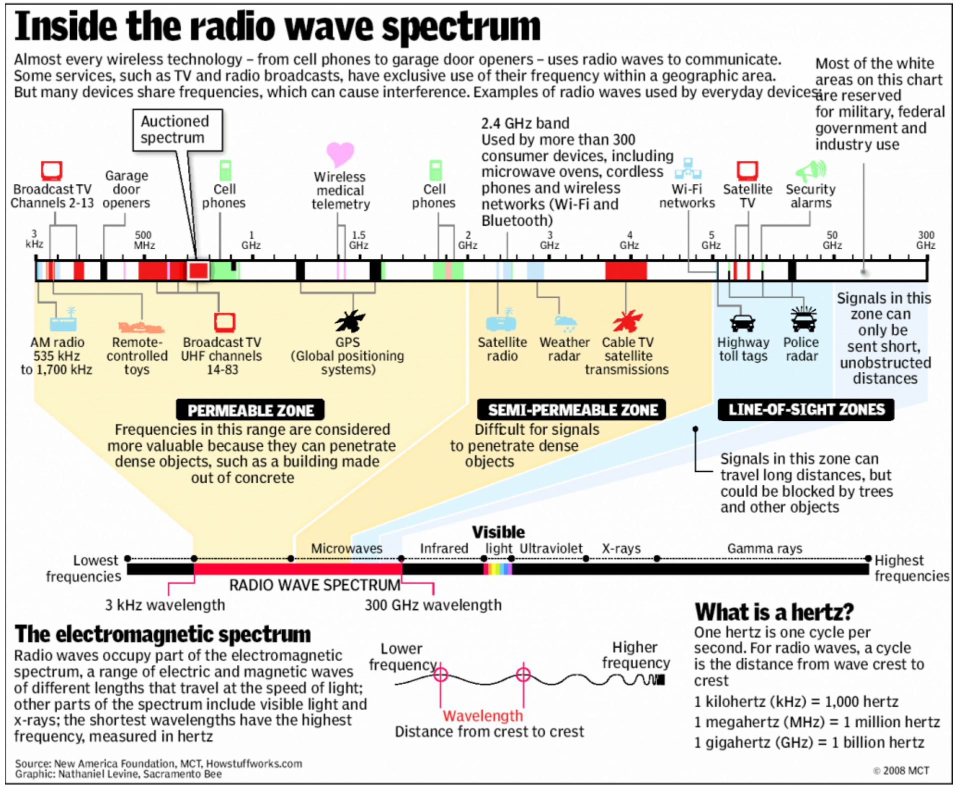


Importance of Bandwidth

- Data rate generally scales linearly in bandwidth
  - If the transmit power and bandwidth increase by $N \Rightarrow$ the communication rate increase by N
  - We will see this in detail later
- Ex: Compare GSM (2G) and LTE (4G)
  - Single channel of GSM system = 200 kHz
  - Single channel of LTE = 20 MHz
  - If power scales sufficiently, LTE would in general have 100x data rate
  - LTE, in fact, can have even more capacity due to other improvements
- Figure to the right: 802.11ad channels
  - The channels are > 2 GHz



Radio Spectrum

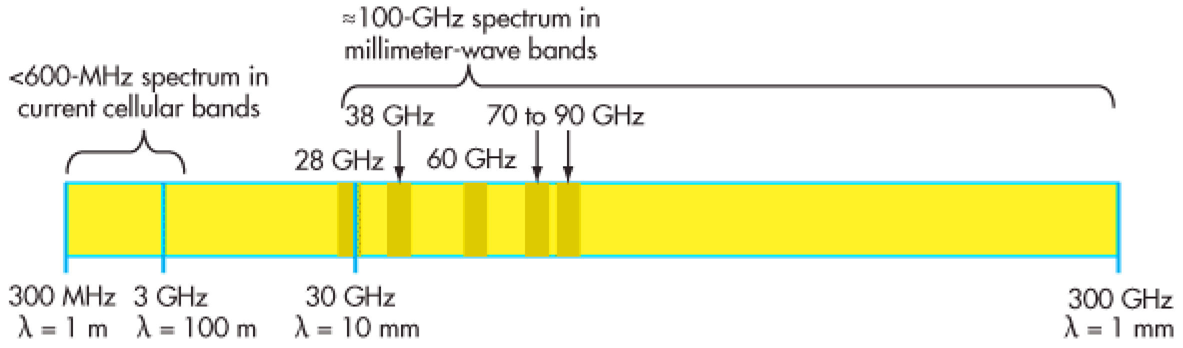


Bandwidth and Center Frequencies Examples

System	Duplex	Center freq (MHz)	Bandwidth
GSM	FDD	GSM-850: 824-849 (UL), 869-894 (DL) GSM-900: 890-914 (UL), 935-959 (DL) GSM-1800: 1710–1784(UL), 1805.2–1879(DL) GSM-1900: 1850–1910(UL), 1930–1990(DL)	200 kHz per channel
UMTS	FDD	GSM + other bands ~2100 and ~1900	5 MHz per carrier
LTE	Mostly FDD	Mostly in 2100 to 2600 MHz	1.4 to 20 MHz, 10 MHz typical
802.11abg	TDD	2.4 GHz (ISM band) and 5 GHz (U-NII band)	20 MHz
802.11n			20, 40 MHz
802.11ac			20-160 MHz
802.11ad	TDD	60 GHz (millimeter wave spectrum)	2.16 GHz

Millimeter Wave Bands

- New bands for 5G
  - 100x more bandwidth than conventional bands below 6 GHz
  - Bands at 28 GHz and 38 GHz opened up by FCC
  - 5G systems operating are in trials now



Outline

- ❑Basics of Electromagnetic Waves
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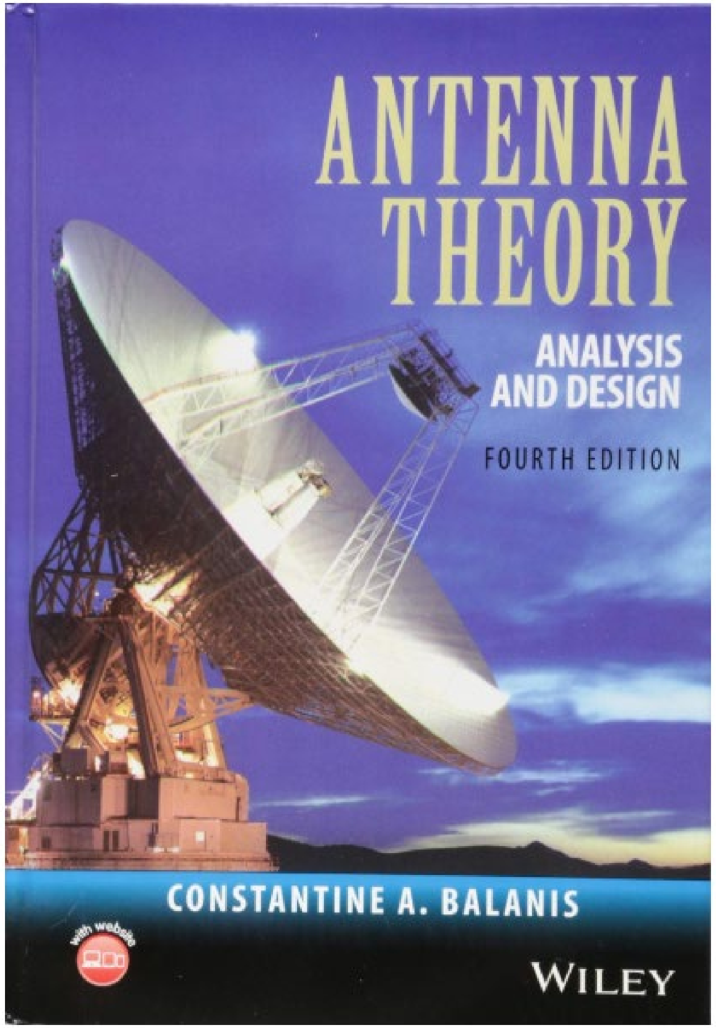
- ❑Basics of Antennas
- ❑Free Space Propagation

Excellent Text for Antennas

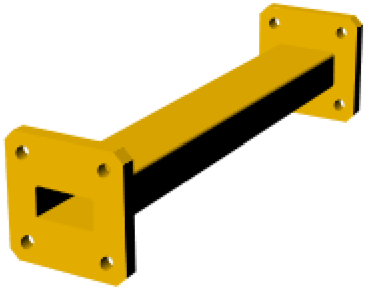
- This lecture is based on classic text
  - Balanis, “Antenna Theory”
  - Most of the figures here are from this text
- If you want to learn more, study the text:
  - Provides full EM theory view
  - Many excellent problems and examples
  - Designed for RF engineers
- We will use only a small portion here
- Take an EM class for more!



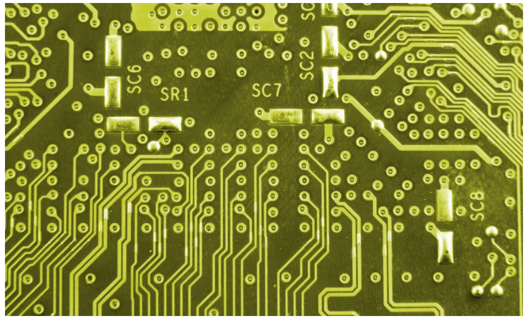
Waveguides and Transmission Lines

- Transmission lines and waveguides: Any structure to guide waves with minimal loss
- Some texts:
  - Transmission lines refer to conductors and waveguides to hollow structures
- Many examples

Coaxial cable



Waveguide



PCB traces
Microstrip: External layer
Stripline: Internal layer

Antenna

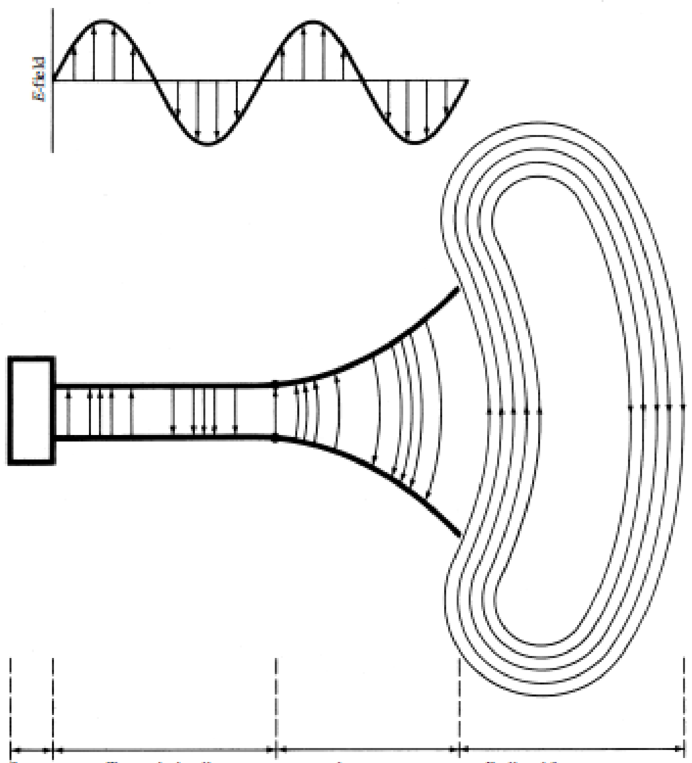


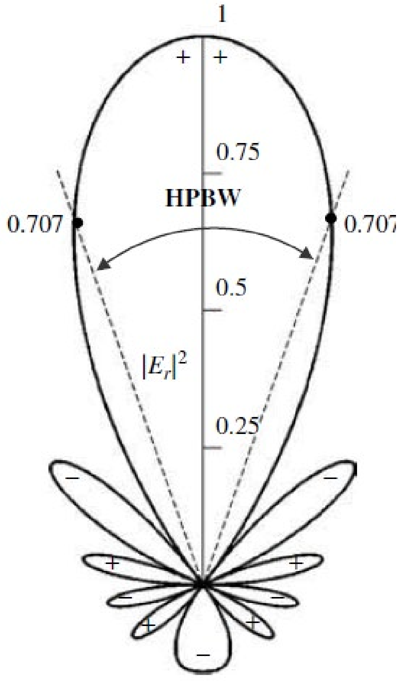
Figure 1.1 Antenna as a transition device.

- Transmit antenna: Radiates electromagnetic waves
- Converts signals:
  - From guided signals in transmission lines to
  - To radiation in free space
- Receive antenna: Collects EM wave

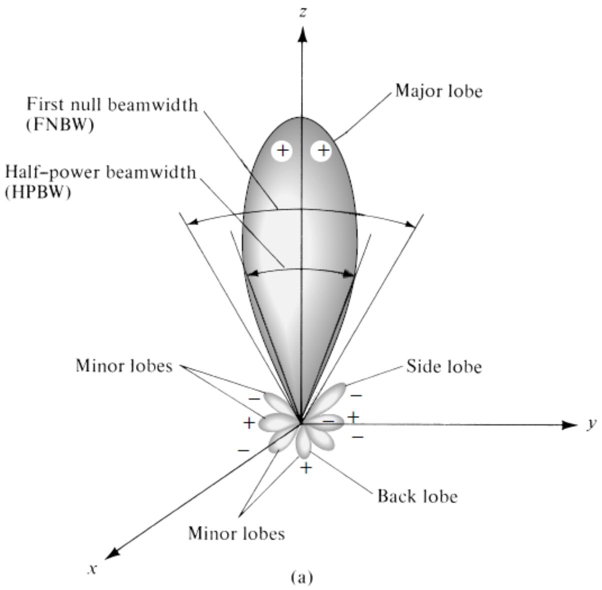
USRP with four vertical antenas

Radiation Patterns

- Aantenna radiation typically shown via a pattern
  - Value of scalar as a function of position
  - Antenna usually at origin
  - Orientation of the antenna is important
- Many possible quantities:
  - Power, electric field, …
  - Normalized or un-normalized
- Can be 2D or 3D



2D

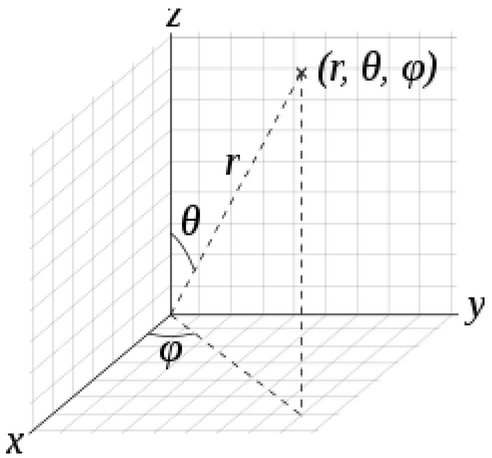


3D

Spherical Coordinates

- Radiation patterns are often given in spherical coordinates
- Spherical coordinates: (φ, θ, r)
  - $\varphi \in [-\pi,\ \pi]$: Azimuth, counter-clockwise angle in xy plane
  - $\theta = \theta_{el} \in [\frac{\pi}{2}\ \frac{\pi}{2}]$: Elevation, angle from xy plane
  - $r \geq 0$: Radius from origin
- Many texts use polar or inclination angle:
  - Use $\theta_{inc} = \frac{\pi}{2} - \theta_{el} \in [0, \pi]$
  - Measures angle from z axis
  - Most antenna and math texts use polar form
  - But, MATLAB antenna toolbox uses elevation form
- Remember right hand rule!

Polar coordinates



Spherical (polar form) ⇔ Cartesian

$$r = \sqrt{x^2 + y^2 + z^2},$$
$$\varphi = \arctan \frac{y}{x},$$
$$\theta = \arccos \frac{z}{\sqrt{x^2 + y^2 + z^2}}$$

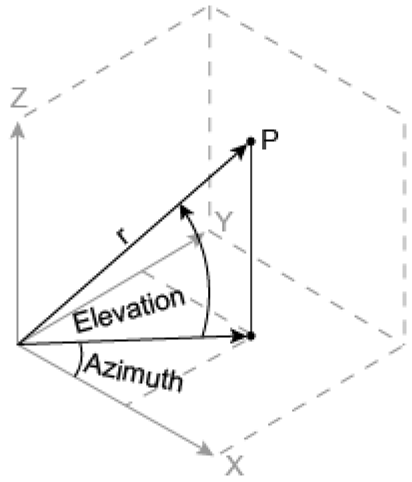
$$x = r \sin\theta \cos\varphi,$$
$$y = r \sin\theta \sin\varphi,$$
$$z = r \cos\theta.$$

Spherical Coordinates in MATLAB

- Conversion between spherical and cartesian

```
% Generate four random points in 3D
X = randn(3,4);

% Compute spherical coordinates of a matrix of points
% Note these are in radians!
[az, el, rad] = cart2sph(X(1,:), X(2,:), X(3,:));

% Convert back
[x,y,z] = sph2cart(az,el,rad);
Xhat = [x; y; z];
```

```
x = r .* cos(elevation) .* cos(azimuth)
y = r .* cos(elevation) .* sin(azimuth)
z = r .* sin(elevation)
```



- Conversion to a coordinate system

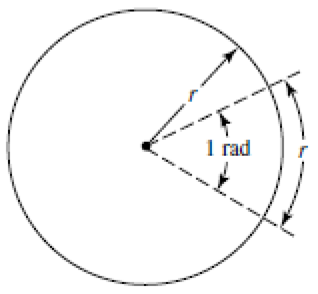
```
%% Conversion to a new frame of reference

% Angles of new frame of reference
% Note these are in degrees!
az1 = 0;
el1 = 45;

% Rotate to the new frame of reference
% This takes row vectors!
X1 = cart2sphvec(X,az1,el1);
```

Radians and Steradians

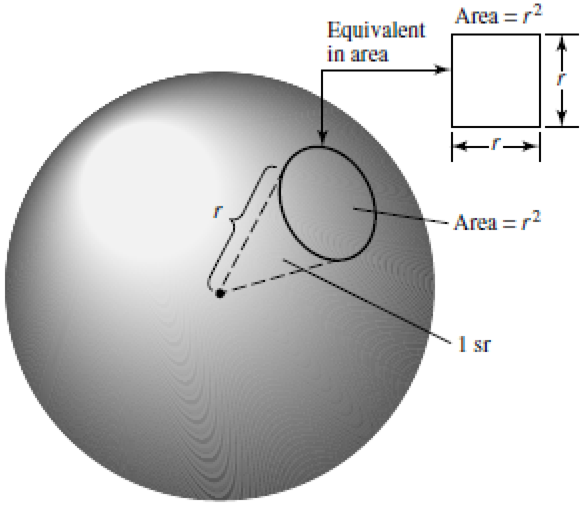
- **Radian**:
  - Circle of radius one
  - Angle for unit length on circumference
  - 2π radians in the circle
- **Steradian**
  - Defined on sphere of radius one
  - Angles corresponding to unit area on surface
  - 4π sr in the sphere
- **Infinitesimal area and solid angle:**

$$dA = r^2 \sin\theta \, d\theta \, d\phi \quad (\mathrm{m}^2) \qquad d\Omega = \frac{dA}{r^2} = \sin\theta \, d\theta \, d\phi \quad (\mathrm{sr})$$

  - Note: θ is the inclination angle not elevation



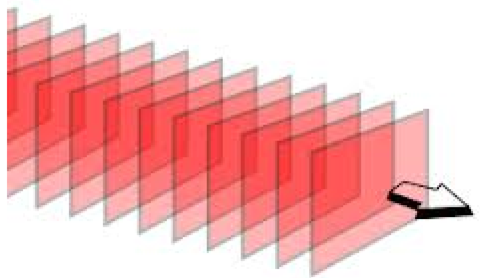
(a) Radian



(b) Steradian

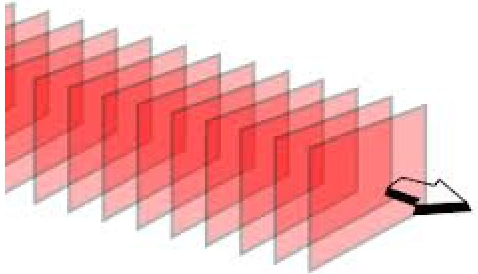
Radiation Density

- Recall instantaneous energy flux for a plane wave: $\boldsymbol{S}(t) = \frac{1}{\mu}\boldsymbol{E}(t) \times \boldsymbol{B}(t) = \frac{1}{\eta}\|\boldsymbol{E}(t)\|^2\boldsymbol{n}$
  - $\boldsymbol{n}$ = normal vector in direction of the plane wave, $\eta = c\mu$ = characteristic impedance
- Typically consider fields at some frequency $\omega = 2\pi f$: $\boldsymbol{E}(t) = Re\left[\boldsymbol{E}e^{i\omega t}\right]$
- Time average power $\langle \boldsymbol{S}(t) \rangle = \frac{1}{2\eta}\|\boldsymbol{E}\|^2\boldsymbol{n}$
  - Note factor of 2
- Can write $\langle \boldsymbol{S}(t) \rangle = W\,\boldsymbol{n}, W = \frac{1}{2\eta}\|\boldsymbol{E}\|^2$
  - Radiation density: $W = W(r,\theta,\phi) = \frac{1}{2\eta}|E(r,\theta,\phi)|^2$ = radiation density
  - Maximum power available if aligned in the direction $\boldsymbol{n}$
  - Units W/m^2
  - This is a function of position $W(r,\theta,\phi)$



Radiation Intensity

- From previous slide: Radiation density: $W = W(r, \theta, \phi) = \frac{1}{2\eta} |\boldsymbol{E}(r, \theta, \phi)|^2$
  - Units $\frac{W}{m^2}$
- Also define radiation intensity: $U = r^2 W = \frac{r^2}{2\eta} |\boldsymbol{E}(r, \theta, \phi)|^2$
  - Watts per solid angle: $\frac{W}{sr}$
- In far field, radiation pattern typically decays as:
  - $\boldsymbol{E}(r, \theta, \phi) \approx \frac{1}{r} \boldsymbol{E}_0(\theta, \phi)$
  - In this case, $U(r, \theta, \phi) = r^2 W(r, \theta, \phi) = \frac{r^2}{2\eta} |\boldsymbol{E}(r, \theta, \phi)|^2 \approx \frac{1}{2\eta} |\boldsymbol{E}_0(\theta, \phi)|^2$
  - Only depends on angular position $U(r, \theta, \phi) = U(\theta, \phi)$
  - Does not depend on distance r



Total Radiated Power

- Total radiated power:

$$P_{rad} = \iint U d\Omega = \int_{-\pi/2}^{\pi/2} \int_{-\pi}^{\pi} U(\theta, \phi) \cos\theta \, d\phi d\theta$$

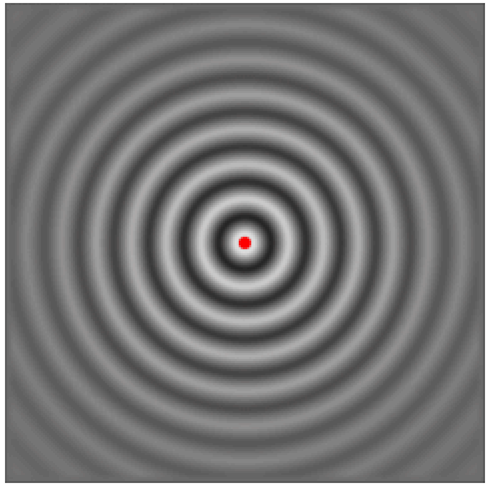
  - Units is Watts
  - Note $\cos\theta$ term! Angle here is elevation angle not polar angle

- Typically measured in dBm or dBW:
  - $P_{rad}[\text{dBm}] = 10 \log_{10} \left[\frac{P_{rad}}{1\,\text{mW}}\right], P_{rad}[\text{dBW}] = 10 \log_{10} \left[\frac{P_{rad}}{1\,\text{W}}\right]$
  - Power relative to mW or W

Isotropic Antenna

- Isotropic antenna: Radiates uniformly in all directions
- Radiation density and intensity are uniform
  - Radiation density: $W(\theta, \phi, r^2) = \frac{P_{rad}}{4\pi r^2}$
  - Radiation intensity: $U(\theta, \phi) = \frac{P_{rad}}{4\pi}$
  - Do not depend on angles θ, ϕ
- Mostly theoretical construct:
  - Most real antennas have some "directivity"
- In fact, there can be no coherent (linearly polarized) isotropic radiator
  - E-field would be always tangent to sphere
  - Such an E-field would have to go to zero in at least one point ("Hairy Ball Theorem")

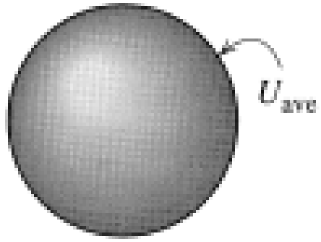
Theoretical isotropic pattern

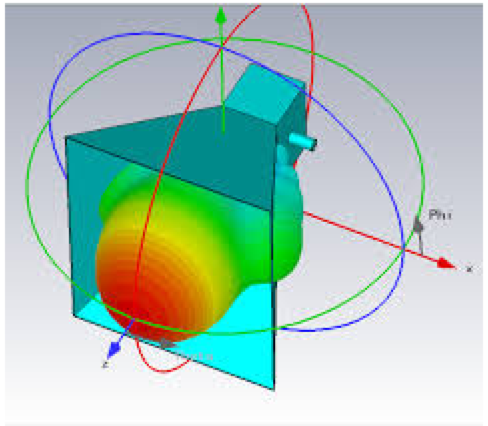


Antenna Directivity

- Most real antennas concentrate power in certain angles
  - They are non-isotropic
- Antenna directivity:
  - $D(\theta,\phi) = \frac{4\pi U(\theta,\phi)}{P_{rad}}$ [dimensionless]
  - Measures power at an angle relative to average
  - Average in linear domain is one
  - For isotropic antenna, $D(\theta,\phi) = 1$
- Max directivity: $D_{max} = \max D(\theta,\phi)$
  - Directivity in direction with maximum power
- Typically measured in dBi
  - dB relative to isotropic
  - $D(\theta,\phi)\ [dBi] = 10 \log\left[\frac{4\pi U(\theta,\phi)}{P_{rad}}\right]$

Theoretical isotropic antenna





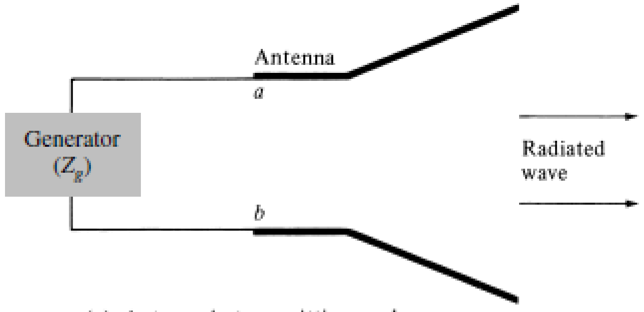
Horn antenna with directivity

Antenna Gain and Efficiency

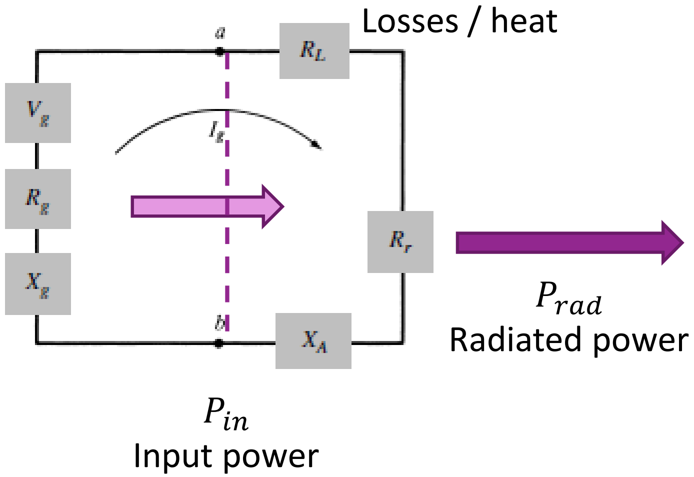
- Most antennas have losses
- Define efficiency:

$$\epsilon = \frac{P_{rad}}{P_{in}} \in [0,1]$$

  - Radiated to input power in TX mode
  - Remaining power is lost in heat in the antenna
  - Losses in the conductor and dielectric
- Lossless antenna: $\epsilon = 1$
- Antenna gain:
  - $G(\theta,\phi) = \epsilon D(\theta,\phi) = \frac{4\pi U(\theta,\phi)}{P_{in}}$
  - Radiation intensity per unit input power
  - For losses antennas, gain = directivity



(a) Antenna in transmitting mode



Antenna Toolbox in MATLAB

- Powerful routines for:
  - Design and analysis of antennas
- Benefits:
  - Supports many antennas
  - Accurate EM modeling
  - Nice visualization tools
  - Simple to use
- Also, free to NYU students
  - Just download it with MATLAB
- But…very slow for complex antennas

Antenna Toolbox

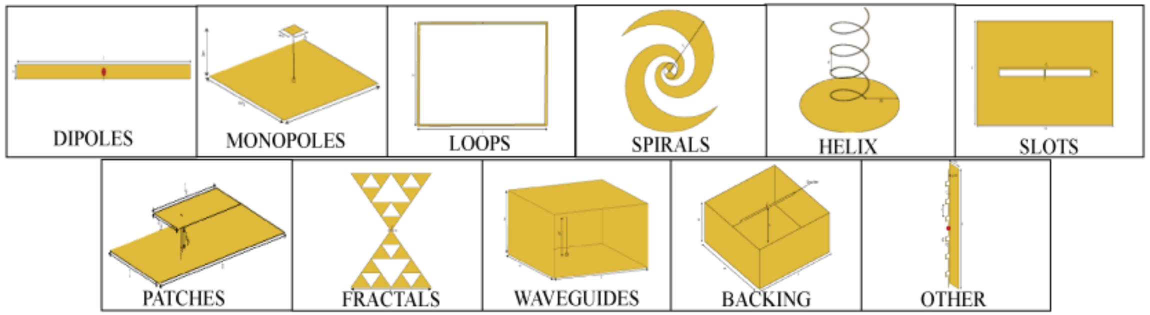
Design, analyze, and visualize antenna elements and antenna arrays

Antenna Toolbox™ provides functions and apps for the design, analysis, and vis
antennas using either predefined elements with parameterized geometry or arbi

Antenna Toolbox uses the method of moments (MoM) to compute port propertie
such as the near-field and far-field radiation pattern. You can visualize antenna g

You can integrate antennas and arrays into wireless systems and use impedanc
beam forming and beam steering algorithms. Gerber files can be generated fron
large platforms such as cars or airplanes and analyze the effects of the structure
using a variety of propagation models.

Get Started
Learn the basics of Antenna Toolbox

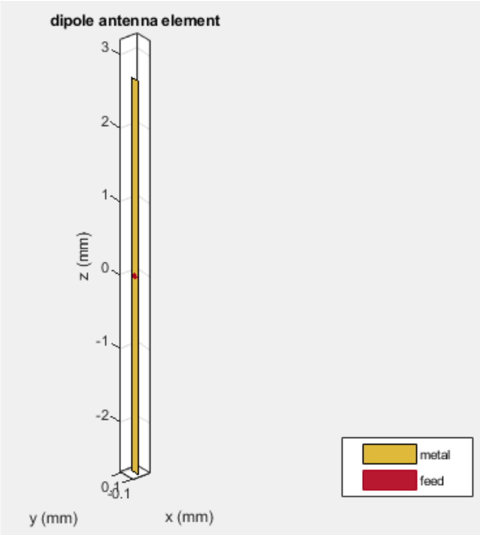


Patterns in MATLAB: Dipole Example

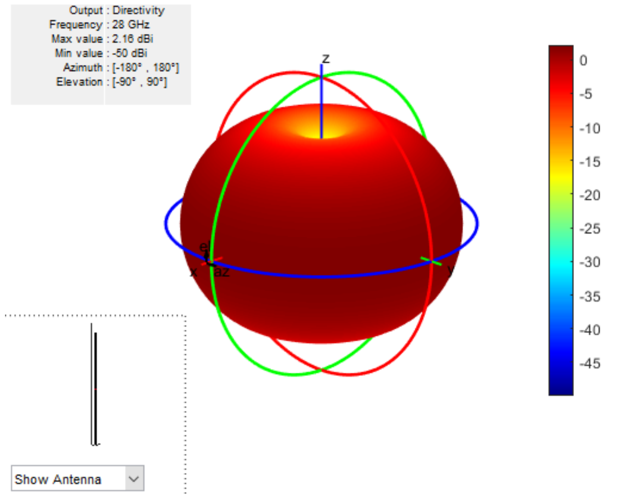
- MATLAB has powerful tools for calculating antenna patterns

```
%% Simulation constants
fc = 28e9;
vp = physconst('lightspeed');
lambda = vp/fc;
```

```
%% Dipole antenna
% Construct the antenna object
ant = dipole(...
    'Length', lambda/2,...
    'Width', 0.01*lambda );
```




```
ant.show();
```



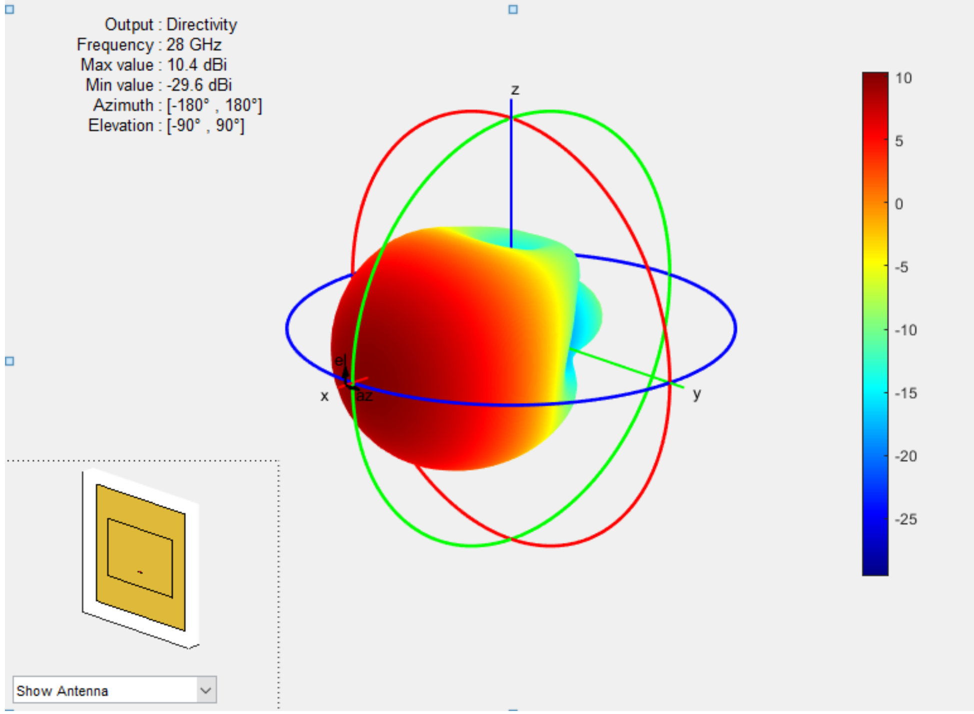

```
ant.pattern(fc)
```

Microstrip Patch Example

- A more complex antenna
- Many other parameters
  - Substrate selection (e.g. FR4, Rogers)
  - Shapes, notches, …

```
%% Create a patch element
len = 0.49*lambda;
groundPlaneLen = lambda;
ant2 = patchMicrostrip(...
    'Length', len, 'Width', 1.5*len, ...
    'GroundPlaneLength', groundPlaneLen, ...
    'GroundPlaneWidth', groundPlaneLen, ...
    'Height', 0.01*lambda, ...
    'FeedOffset', [0.25*len 0]);
```

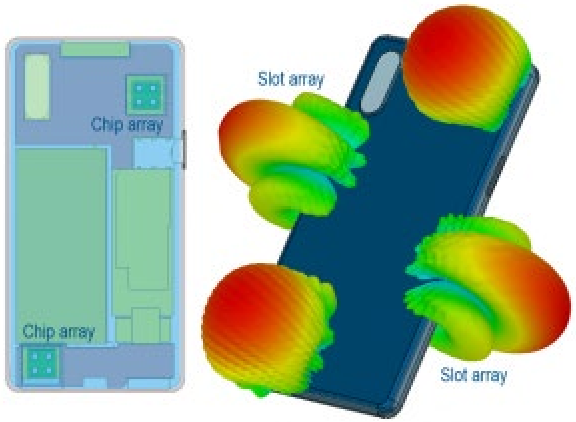
```
%%
% Tilt the element so that the maximum energy is in the x-axis
ant2.Tilt = 90;
ant2.TiltAxis = [0 1 0];

% Display the antenna pattern after rotation
ant2.pattern(fc);
```



More Complex Antennas

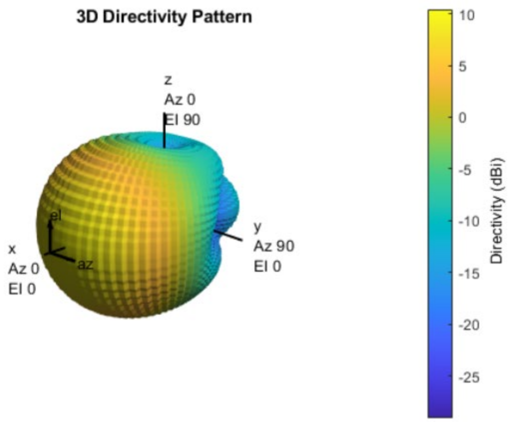
- For complex antennas:
  - MATLAB antenna toolbox is often too slow
  - Cannot handle packaging, covers, obstacles, …
  - Need other tools (e.g. Ansoft HFSS and CST)
- Use MATLAB custom antenna object
  - Store offline computed pattern

```
phasePattern = zeros(size(dir));
ant3 = phased.CustomAntennaElement(...
    'AzimuthAngles', az, 'ElevationAngles', el, ...
    'MagnitudePattern', dir, ...
    'PhasePattern', phasePattern);

% Plot the antenna pattern.
% Note the format is slightly different since we are using
% the pattern routine from the phased array toolbox
ant3.pattern(fc);
```



CST simulation of 28 GHz array on a handset with cover

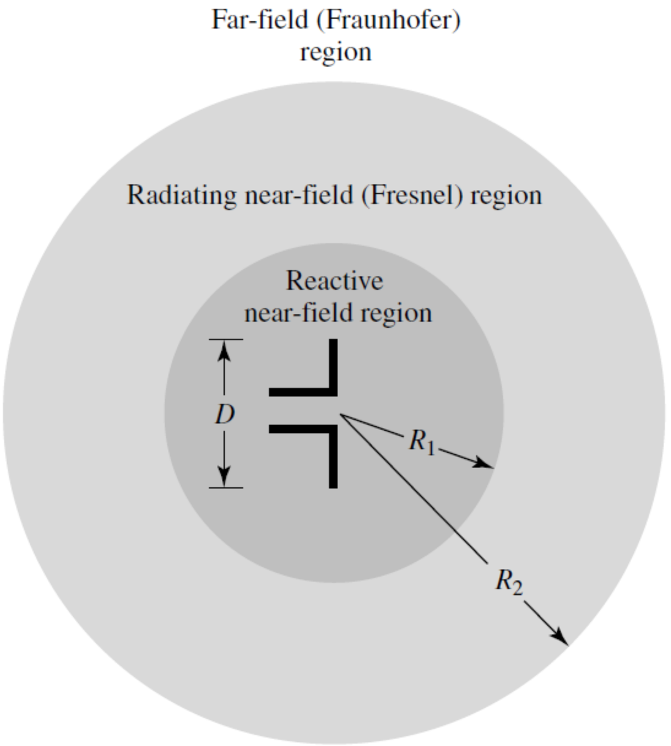
https://blogs.3ds.com/simulia/5g-antenna-design-mobile-phones/



Demo of custom antenna element in MATLAB

Field Regions

- ❑ Antenna patterns depend on the region
- ❑ Reactive near field:
  - Reactive pattern dominates
- ❑ Radiating near field or Fresnel region:
  - Angular pattern depends on distance
- ❑ Far field or Fraunhofer region:
  - Angular pattern independent of distance
  - Radiation is approximately plane waves
- ❑ Can be approximately calculated using:
  - D: Maximum antenna dimension
  - λ: Wavelength

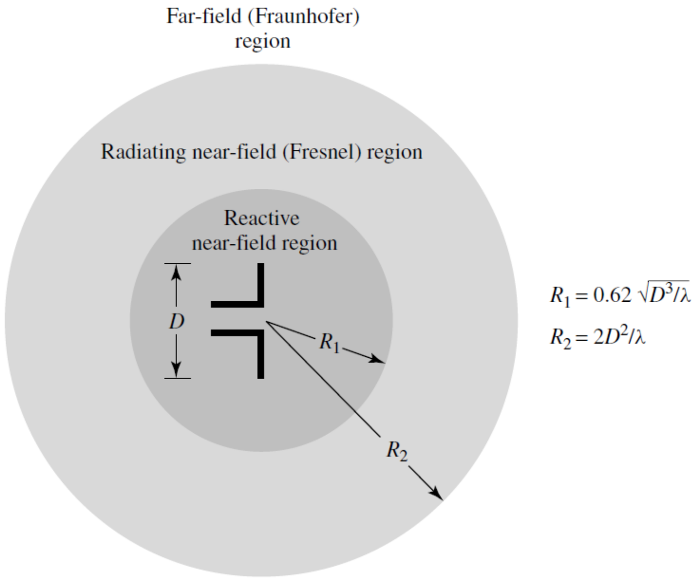


$R_1 = 0.62\sqrt{D^3/\lambda}$

$R_2 = 2D^2/\lambda$

Rayleigh Distance

- Distance R_2 to far-field = Rayleigh distance
- Most cellular / WLAN systems operate in far field
- Ex 1: Half wavelength dipole antenna
  - f_c = 2.3 GHz:
  - $D = \frac{\lambda}{2}, \; R_2 = \frac{2D^2}{\lambda} = \frac{\lambda}{2} = 6.5 \text{ cm}$
- Ex 2: Large cellular base station
  - $D \approx 7\text{m}$, f_c = 2.3 GHz
  - $R_2 = 751 \text{ m}$
- Ex 3: MmWave wide aperture antenna
  - $D \approx 40 \text{ cm}$, f_c = 140 GHz
  - $R_2 = 149 \text{ m}$



Outline

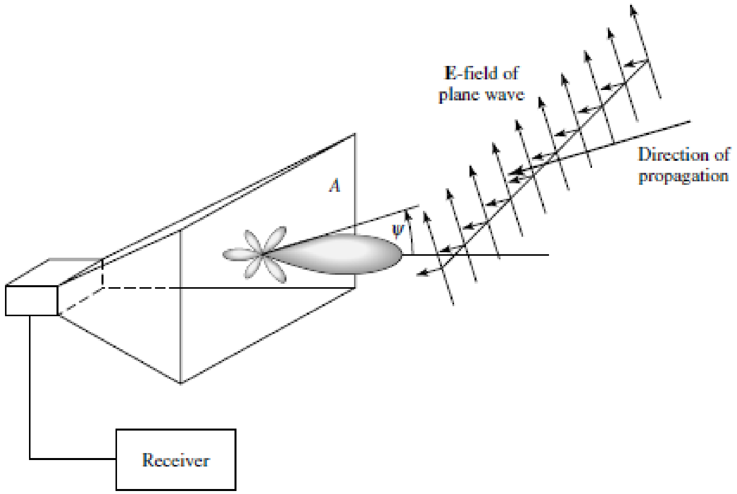
- Basics of Electromagnetic Waves
- Power and Bandwidth of Signals
- Basics of Antennas
- Free Space Propagation

Antenna Effective Aperture

- Suppose RX antenna sees incident plane wave
  - Assume polarization aligned to the antenna
- The effective antenna aperture (or area):

$$A_e(\theta,\phi) = \frac{W(\theta,\phi)}{P_L} \quad [m^2]$$

  - W = Power density of incident wave [W / m^2]
  - P_L = Power delivered to load at the receiver [W]
- The effective area that the antenna collects
  - We will see this is different than the physical aperture
- A_e will depend on the direction of arrival

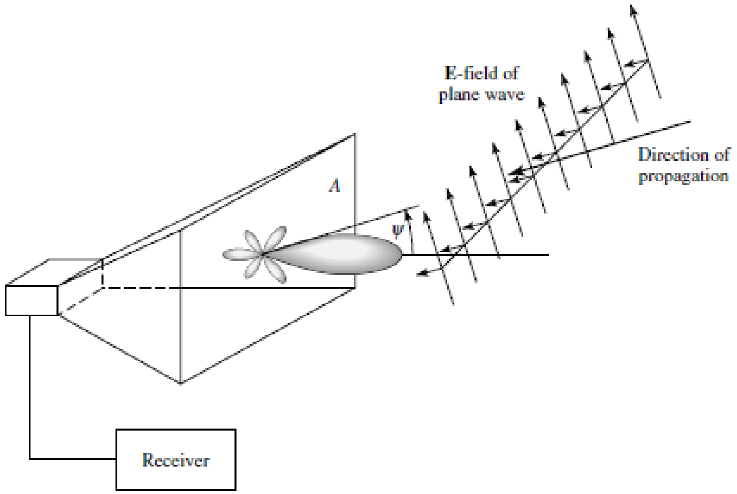


Aperture and Directivity

- From previous slide, effective aperture is: $A_e(\theta,\phi) = \frac{W(\theta,\phi)}{P_L}$ $[m^2]$
  - Ratio of received power to incident radiation density
- Aperture-directivity relation:

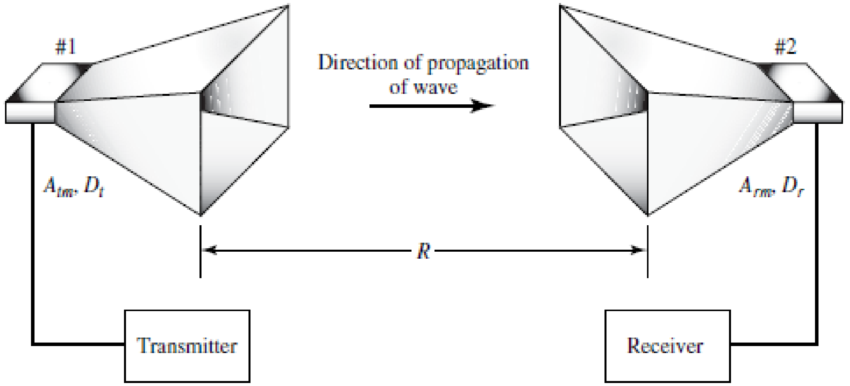
$$A_e(\theta,\phi) = D(\theta,\phi)\,\frac{\lambda^2}{4\pi}$$

  - True for all lossless antennas
  - Proof: next slide
- Consequence: Average aperture is always $\frac{\lambda^2}{4\pi}$
  - Why? $\frac{1}{4\pi}\iint A_e(\theta,\phi)\cos\theta\, d\theta d\phi = \frac{\lambda^2}{(4\pi)^2}\iint D(\theta,\phi)\cos\theta\, d\theta d\phi = \frac{\lambda^2}{4\pi}$
- Independent of the physical size of the antenna!



Proof of the Aperture-Directivity Relation

- Suppose Ant 1 transmits power P_t
- Radiation density is: $W = \frac{D_1 P_t}{4\pi R^2}$
- Received power at Ant 2: $P_r = A_2 W = \frac{A_2 D_1 P_t}{4\pi R^2} \Rightarrow \frac{P_r}{P_t} = \frac{A_2 D_1}{4\pi R^2}$
- TX from Ant 2, the gain must be the same: $\frac{P_r}{P_t} = \frac{A_1 D_2}{4\pi R^2}$
  - This is a consequence of reciprocity
- Hence, for *any* two antennas: $\frac{D_1}{A_1} = \frac{D_2}{A_2}$
- From simple antenna calculations for a short dipole:
  - $D_2 = \frac{3}{2},\ A_2 = \frac{3\lambda^2}{8\pi} \Rightarrow \frac{D_2}{A_2} = \frac{4\pi}{\lambda^2}$ (Needs basic EM theory)

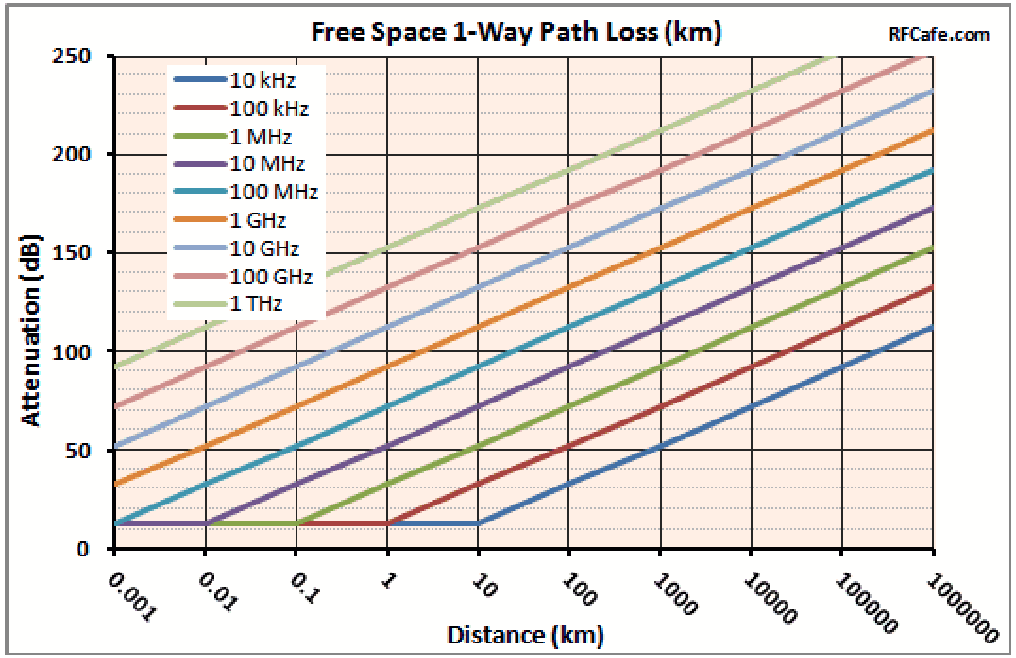


Friis' Law

- Consider two lossless antennas in free space
- From previous slide: $\frac{P_r}{P_t} = \frac{A_1 D_2}{4\pi R^2}$
- From aperture-directivity relation: $A_1 = D_1 \frac{\lambda^2}{4\pi}$
- This leads to Friis' Law (for lossless antennas):

$$\frac{P_r}{P_t} = D_1 D_2 \left(\frac{\lambda}{4\pi R}\right)^2$$

  - Path loss is proportional to R^2
  - Path loss Inversely proportional to $\lambda^2 \Rightarrow$ proportional to f_c^2

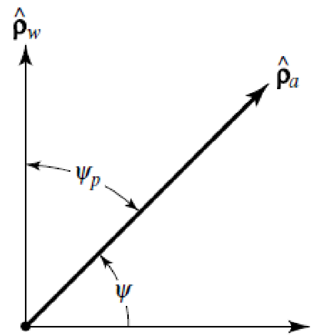


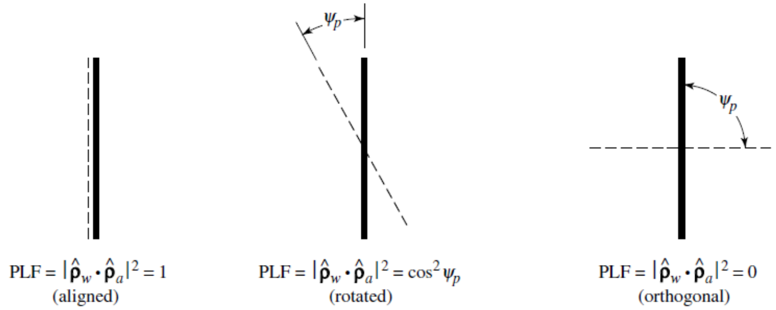
Polarization Loss

- ❑ Friis' Law assumes incident wave is aligned in polarization
- ❑ In general, need to consider polarization loss
- ❑ Recall: polarization vector for a plane wave:
  - Direction of the E-field in phasor notation
  - A complex vector in 3-dim
- ❑ Polarization loss factor:

$$PLF = |\boldsymbol{\rho}_a \cdot \boldsymbol{\rho}_w|^2 = \cos^2 \psi_p$$

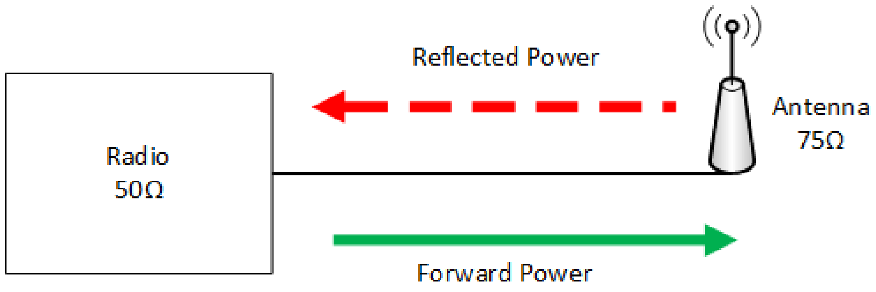
  - $\boldsymbol{\rho}_a$: Polarization vector of the TX wave from antenna
  - $\boldsymbol{\rho}_w$: Polarization vector of the RX incident wave
  - ψ_p: Angle between them

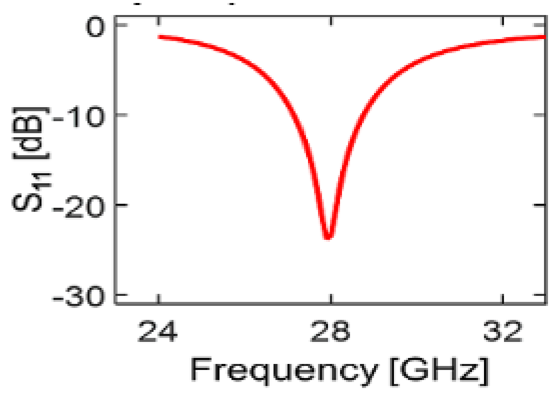




Antenna Impedance and Matching

- Not all power from radio may be delivered to antenna
- Some is reflected back
- Described by reflection coefficient Γ
  - Also referred to as S_{11}
  - Complex ratio of forward to reverse wave
- Also described by impedance mismatch:
  - $\Gamma = \frac{Z_{in} - Z_0}{Z_{in} + Z_0}$
- Fraction of power transferred: $1 - |\Gamma|^2$
- Also given as voltage standing wave ratio (VSWR) $= \frac{1+|\Gamma|}{1-|\Gamma|}$

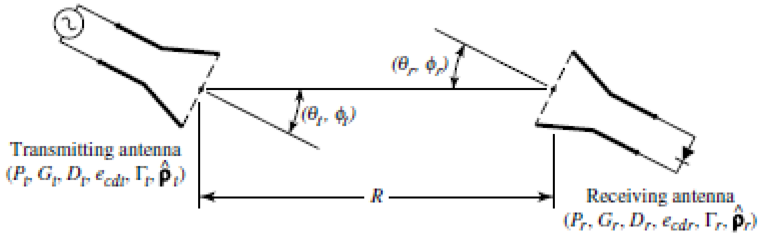




Ali et al, Small Form Factor PIFA Antenna Design at 28 GHz for 5G Applications, 2019

Friis' Law with Losses

- Three losses in practice:
  - Polarization loss
  - Conductive / dielectric loss
  - Impedance mismatch



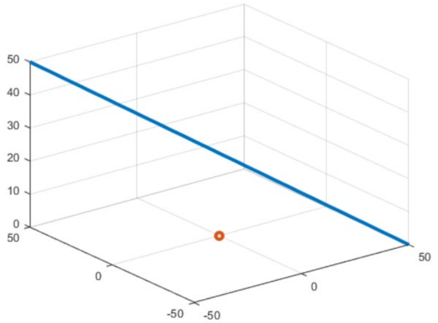
- Friis' Law with lossy antennas:

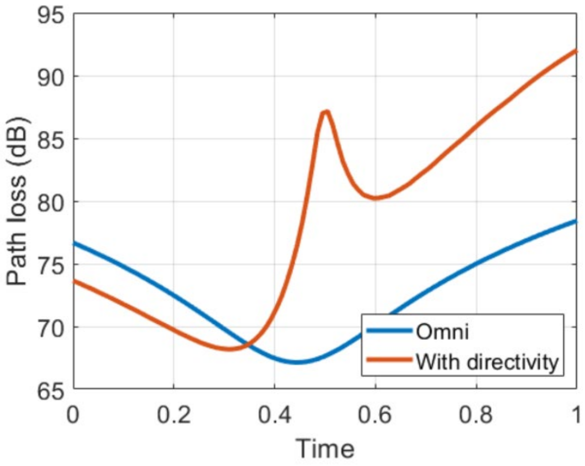
$$\frac{P_r}{P_t} = \epsilon_1 \epsilon_2 (1 - |\Gamma_1|^2)(1 - |\Gamma_2|^2) D_1 D_2 \left(\frac{\lambda}{4\pi R}\right)^2 \cos^2 \theta_{POL}$$

  - ϵ_i: Efficiency of antenna
  - θ_{POL}: Angle between the polarization vectors
  - Note that gain is: $G_i = \epsilon_i D_i$

Demo: Plotting Path Loss on a Path

- Compute the path loss on a path
  - TX is isotropic at the origin
  - RX is micro-strip patch element. Moves along a line
- MATLAB code
  - Computes angles to RX
  - Computes free-space omni directional path loss
  - Interpolates directivity
  - Adds directivity to FSPL






```
[azpath, elpath, dist] = cart2sph(X(1,:), X(2,:), X(3,:));
azpath = rad2deg(azpath);
elpath = rad2deg(elpath);

% Compute the free space path loss along the path without
% the antenna gain.  We can use MATLAB's built-in function
plOmni = fspl(dist, lambda);

% Compute the directivity using interpolation of the pattern.
% We can use the  |ant3.resp| method for this purpose, but the
% interpolation is not smooth.  So, we will do this by hand using
% MATLAB's interpolation objects.
F = griddedInterpolant({el,az},dir);

% Compute the directivity using interpolation
dirPath = F(elpath,azpath);

% Compute the total path loss including the directivity
plDir = plOmni - dirPath;

% Plot the path loss over time.  Can you explain the
plot(t, [plOmni; plDir]', 'Linewidth', 3);
grid();
set(gca, 'Fontsize', 16);
legend('Omni', 'With directivity', 'Location', 'SouthEast');
xlabel('Time');
ylabel('Path loss (dB)');
```